

TRANSFORMER LAMINATIONS, DESIGN CONSIDERATIONS

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Engineers and designers of transformers and inductors will find the information presented in the following helpful in the selection of magnetic materials and core shapes for specific applications.

Why Magnetic Materials

Magnetic materials are useful for the generation and distribution of electrical power because these materials allow to transmit large power densities at low losses. In addition, voltage and impedance can be easily changed from one level to another, since changes in the flux density of materials induce voltages in copper coils surrounding the magnetic cores (Faraday's law).

The energy density in a magnetic material is

$$E = \frac{1}{2}HB \left(\frac{A}{\text{cm}} \frac{V}{\text{cm}^2} \right) \text{ or } \left(\frac{\text{VAs}}{\text{cm}^3} \right) \quad (1)$$

H is the magnetic field, B the magnetic induction (1 Vs/cm² = 10⁸ Gauss = 10⁴ Tesla). In a field of 500 A turns/cm and an induction of 2 Tesla, an energy density of 5 · 10⁻² W/cm³ can be stored, which is as high as in the best capacitors. By multiplying equation (1) with core volume V_c = A_c · lm, where A_c is the core cross section and lm the mean path length and assuming a sinusoidal change of B at the frequency f, it can be rewritten as follows. The power handling capacity in VA is

$$VA = 4.44 \text{ lm } A_c f B H 10^{-8} \quad (2)$$

Since H = ni/lm, n number of turns, i current in the turns and ni = S Aw K, S current density in the copper wire, Aw core window, 2K copper fill factor (.35% for primary and secondary turns), the power handling capacity of a transformer as derived from the energy storage equation is

$$VA = 4.55 S B f A_c A_w \cdot 10^{-4} \quad (3)$$

in which A_c and A_w are in in², B in Tesla, S in A/in² and f is the frequency.

The same result is, of course, obtained if we multiply Faraday's law for induction with the current i. E_i = -n A_c dB/dt, where E_i is a voltage induced in n turns by a flux change dB/dt. Solving this equation for

sinusoidal flux, we find

$$E_i = 4.44 n f B A_c \cdot 10^{-4} \text{ volts,}$$

in which A_c is in square cm multiplying with i, we find for the power VA

$$VA = 4.44 n f B A_c i \cdot 10^{-4} \quad (4),$$

since ni = S · Aw · K, we can substitute in equation 4 and transform into in², so that VA = 4.55 S B · f A_c A_w · 10⁻⁴, (5) = (3) in volt amperes.

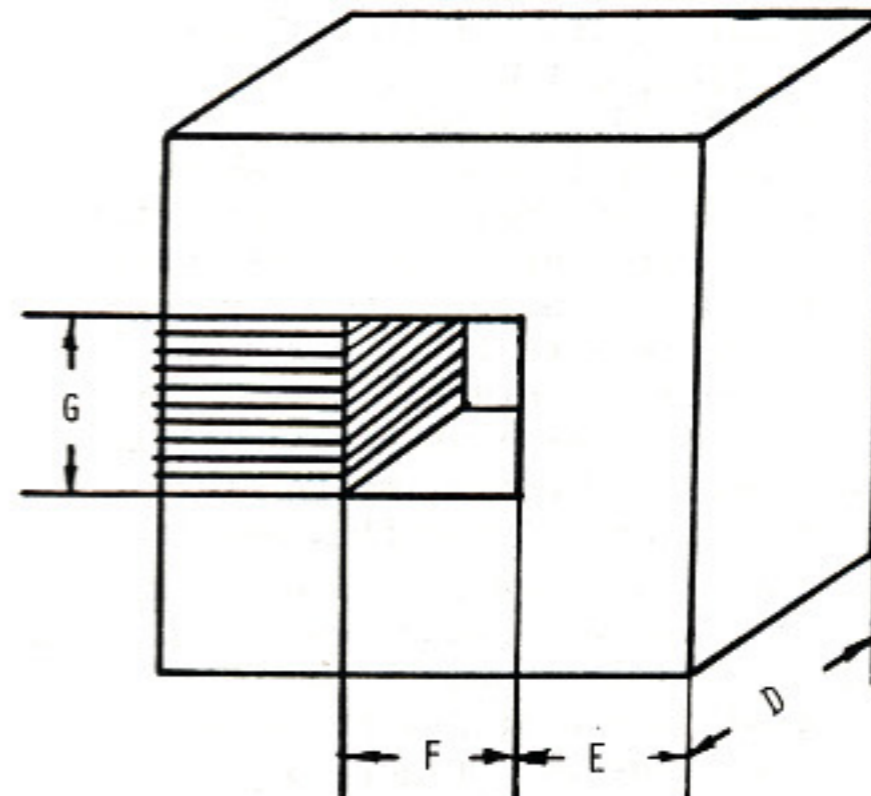


Fig. 1 Transformer Core

A_c = ED = cross section of core
A_w = Gf = window area

Metallic magnetic materials can be used from low frequencies of a few Hz to high frequencies of few hundred kHz, ferrites and iron powder cores can be used up into the MHz range. With above equations, the designer can select suitable dimensions for the copper coil and the magnetic core cross section at the given frequency which meets the loss requirement. Most manufacturers of core components list in their catalogs the A_w A_c products for available shapes of core structures. Transformers and inductors can be reduced in weight and volume by operating at higher frequency or by selecting materials which can work at a higher flux density.